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TABLE OF CONTENTS

1. ACTIVE BATTERY BALANCING VIA A SWITCHED DC/DC CONVERTER: DESCRIPTION AND PERFORMANCE ANALYSIS <i>David García Elvira, Hugo Valderrama Blaví, Josep Maria Bosque Moncusí, Ángel Cid Pastor, Juan Antonio Garriga Castillo, Luis Martínez Salamero</i>	1
2. CASCADED OBSERVERS FOR THE ROTOR POSITION, SPEED AND LOAD TORQUE OF THE PMSM USING AN EXTENDED DYNAMIC MODEL <i>Mihai Comanescu</i>	7
3. INFLUENCES OF JOINING PROCESS PARAMETERS ON STATOR AND ROTOR CORES FOR ELECTRIC DRIVES <i>Marco Ziegler, Michael Biburger, Alexander Kuehl, Joerg Franke</i>	11
4. INTELLIGENT CONTROL OF A WIND TURBINE BASED ON REINFORCEMENT LEARNING <i>Nikita Tomin, Victor Kurbatsky, Huseyngulu Guliev</i>	17
5. SHIP POWER SYSTEM QUALITY STUDY INCLUDING PERMANENT MAGNET SYNCHRONOUS MOTOR AND INDUCTION MOTOR DRIVES <i>Nikolay Djagarov, Milen Bonev, Zhivko Grozdev, Hristo Milushev, Julia Djagarova, Dimitar Tsvetanov, Georgi Enchev</i>	23
6. SHIP'S INDUCTION MOTORS FAULT DIAGNOSIS <i>Nikolay Djagarov, Zhivko Grozdev, Georgi Enchev, Julia Djagarova</i>	29
7. ELECTRICAL COIL THERMAL PARAMETERS DETERMINATION BY EXPRESS EXPERIMENTAL DATA PROCESSING <i>Denitsa Darzhanova, Zaharinka Gergova, Petar Darjanov</i>	33
8. IRON-CORED ELECTRICAL WINDING EQUIVALENT THERMAL PARAMETERS INVESTIGATION FOR TRANSIENT RESPONSE SIMULATION <i>Petar Darjanov, Denitsa Darzhanova, Zaharinka Gergova</i>	37
9. MODIFIED LOW-FREQUENCY FILTER FOR SIGNAL SPECTRUM LIMITATION AT ANALOG-DIGITAL CONVERTER'S INPUT WITH DIFFERENTIAL INPUT <i>Darya Yu. Denisenko, Evgeniy A. Zhebrun, Nikolay N. Prokopenko, Yuriy Iv. Ivanov</i>	41
10. BASIC FUNCTIONAL CJFET OP-AMP NODES FOR OPERATION AT LOW TEMPERATURES AND AT CONDITIONS OF PENETRATING RADIATION <i>Anna V. Bugakova, Darya Yu. Denisenko, Nikolay N. Prokopenko, Alexey E. Titov</i>	45
11. STUDY OF THE IMPACT OF INDUCTION MOTORS' STARTING ON THE SUPPLY VOLTAGE <i>Nikolay Matanov</i>	49
12. SIMULATION BASED APPROACH FOR DETERMINING THE BATTERY CONTROL STRATEGY OF A PV AND BATTERY PILOT SYSTEM <i>Vladimir Gjorgievski, Kiril Demerdziev, Bodan Velkovski, Vladimir Shokarovski, Dimitar Dimitrov, Sanja Veleva, Marija Kacarska</i>	54
13. IMPLICATIONS OF RESIDENTIAL BATTERY CHARGE AND DISCHARGE RATES ON SELF-CONSUMPTION AND PEAK POWER EXCHANGE <i>Vladimir Gjorgievski, Snezana Cundeva</i>	59
14. OPTIMIZATION OF WIND-DIESEL POWER PLANTS PARAMETERS AND PLACEMENT FOR POWER SUPPLY OF RUSSIA'S NORTHERN REGIONS CONSUMERS <i>Viktor Elistratov, Inna Bogun, Valeriia Kasina</i>	64

15. OPTIMIZATION OF HYBRID SYSTEMS' OPERATING MODES BASED ON RENEWABLE ENERGY <i>Viktor Elistratov, Roman Denisov</i>	69
16. ANALYSIS OF THE INTERRUPTIONS IN A SECTION OF POWER DISTRIBUTION NETWORK MEDIUM VOLTAGE 20 KV <i>Angel Petleshkov, Yavor Lozanov</i>	73
17. THERMOGRAPHIC STUDY OF THE HEATING OF A PULSE-CONTROLLED MAGNETIC-TECHNOLOGICAL DEVICE <i>Angel Petleshkov, Svetlana Tzvetkova, Yavor Lozanov</i>	78
18. DISTRIBUTED GENERATION CONTROL IN ENERGY MARKET CONDITIONS USING ADVANCED MODEL PREDICTIVE CONTROL <i>Valeriya Dubaylova, Nikita Mukhlynin, Olga Polyakova</i>	82
19. COMPUTATIONAL PERFORMANCE COMPARING OF THE STATE ESTIMATION PROBLEM STATEMENTES IN POLAR AND RECTANGULAR COORDINATES <i>Ilya Polyakov, Andrey Pazderin, Olga Polyakova</i>	86
20. IMPROVING OF THE COMBINED CYCLE POWER PLANT OPERATION UNDER ISLANDING CONDITIONS THROUGH VOLTAGE CONTROL <i>Olga Polyakova, Ildar Gabdulin, Pavel Chusovitin, Andrey Pazderin</i>	90
21. A STRONG ACTION POWER SYSTEM STABILIZER APPLICATION IN A MULTI-MACHINE POWER SYSTEM CONTAINING A NUCLEAR POWER PLANT <i>Aya Abou-El-Soud, Sayed H. A. Elbanna, Waheed Sabry</i>	94
22. VIRTUAL POWER PLANTS MODELING AND SIMULATION USING INNOVATIVE ELECTRO-ECONOMICAL CONCEPT <i>Fatma El Zahraa Magdy, Doaa Khalil Ibrahim, Waheed Sabry</i>	100
23. INNOVATIVE WIND-CUBE TRACKING SYSTEM USING VANE CONTROL: DESIGN AND IMPLEMENTATION <i>Reda Y. Abdellateef, Amal Farouk Abdelgawad, Waheed Sabry, Usama Abouzayed</i>	105
24. AN APPROACH FOR MODELING THE ELECTRONIC CONVERTER-MOTOR SYSTEM FOR ELECTRIC VEHICLES <i>Zahari Zarkov, Vladimir Lazarov, Plamen Rizov, Ludmil Stoyanov, Encho Popov</i>	110
25. IMPEDANCE MATCHING FOR AN INDUCTIVE HARVESTING TRANSDUCER <i>Juan A. Garriga, Hugo Valderrama-Blavi, David García-Elvira, J.A. Barrado, Angel Cid-Pastor</i>	116
26. INVESTIGATION OF TRACTION MOTOR DYNAMIC USING COMPUTER SIMULATION AND METHOD OF MUTUAL LOADING OF TWO PAIR MOTORS <i>Ilya Dvornikovs, Marks Marinbahs, Jaroslavs Zarembo, Edgars Groza, Karlis Ketners</i>	122
27. INVESTIGATION OF ELECTRICAL BUS TRACTION MOTOR DYNAMIC USING METHODS OF PHYSICAL AND COMPUTER SIMULATION <i>Olegs Sliskis, Ilya Dvornikovs, Marks Marinbahs, Janis Marks, Edgars Groza</i>	126
28. SIMO SYSTEM IDENTIFICATION OF TRANSFER FUNCTION MODEL FOR ELECTROHYDRAULIC POWER STEERING <i>Alexander Mitov, Jordan Krlev, Tsonyo Slavov, Ilcho Angelov</i>	130
29. IMPLEMENTATION OF EMBEDDED THREE-LOOP CONTROL ALGORITHM FOR ELECTROHYDRAULIC POWER STEERING SYSTEM <i>Alexander Mitov</i>	136

30. SIMULATION OF THE LOW-FREQUENCY ELECTROMAGNETIC TRANSDUCER OF THE COMPENSATING SOURCE OF THE ACTIVE VIBRATION DAMPING SYSTEM <i>Vitaly Ryzhov, Oleg Molokanov, Pavel Dergachev, Sergey Osipkin, Ekaterina Kurbatova, Pavel Kurbatov</i>	142
31. OPTIMIZED POWER FLOW CONTROL OF SMART GRIDS WITH ELECTRIC VEHICLES AND DER <i>Metody Georgiev, Rad Stanev, Anastassia Krusteva</i>	147
32. A SURVEY OF PERIODIC BINARY NEARLY PERFECT SIGNALS WITH LENGTHS $N=1\text{MOD}4$ <i>Mihail Iliev, Borislav Bedzhev, Monika Bedzheva, Kaloyan Kanchev</i>	153
33. AN ALGORITHM FOR SYNTHESIS OF MISMATCHED FILTERS FOR PROCESSING OF APERIODIC PHASE MANIPULATED SIGNALS <i>Mihail Iliev, Borislav Bedzhev, Monika Bedzheva, Kaloyan Kanchev</i>	159
34. IMPACT OF CAPACITOR ON OPERATING CHARACTERISTICS OF SINGLE PHASE MOTOR <i>Vasilija Sarac, Neven Trajchevski</i>	164
35. REVISITING THE PLOTTING OF MECHANICAL CHARACTERISTIC OF INDUCTION MOTOR WITH A WOUND ROTOR <i>Vasiliy Frizen, Sergey Nazarov, Viktor Denisenko</i>	169
36. RESEARCH OF THE PROCESS HEATING OF DISPERSED STEEL LOAD <i>Aleksandr Antipin, Vasiliy Frizen</i>	173
37. DETERMINATION OF PARAMETERS OF HIGH-FREQUENCY MATCHING TRANSFORMER OF INSTALLATION OF INDUCTION HEATING <i>Aleksandr Antipin, Vasiliy Frizen, Sergey Bychkov</i>	178
38. STUDY OF CHARACTERISTICS OF VIBRATION SPARKING IN HV ROTATING MACHINE INSULATION <i>Victor Belko, Andrey Plotnikov, Yury Petrenya, Tatiana Shikova</i>	182
39. ASPECTS REGARDING TESTS OF THREE-PHASE ASYNCHRONOUS MOTORS WITH SINGLE - PHASE SUPPLY <i>Sorin Enache, Aurel Campeanu, Ion Vlad, Monica-Adela Enache</i>	186
40. PARTICULARITIES REGARDING DESIGN OF ASYNCHRONOUS MOTORS USED FOR DRIVING COAL MILLS <i>Sorin Enache, Aurel Campeanu, Ion Vlad, Monica Adela Enache</i>	192
41. INFLUENCE OF SKIN EFFECT ON STATOR WINDINGS RESISTANCE OF AC MACHINES FOR ELECTRIC DRIVES <i>Ludmil Stoyanov, Vladimir Lazarov, Zahari Zarkov, Encho Popov</i>	198
42. MODELING OF ELECTRICAL CHARACTERISTICS OF VARIOUS PV PANELS <i>Valentin Milenov, Zahari Zarkov, Boris Demirkov, Ivan Bachev</i>	204
43. STUDY OF PHYSICAL MODEL OF WECS WITH SYNCHRONOUS GENERATOR AND BACK-TO-BACK CONVERTER <i>Boris Demirkov, Zahari Zarkov</i>	209
44. PARAMETER ESTIMATION OF AN INTERIOR PERMANENT MAGNET SYNCHROUNOUS MOTOR <i>Emil Sokolov, Miho Mihov</i>	215
45. ANALYSIS OF NON-UNIFORM MAGNETIC FLUX DISTRIBUTION IN CONICAL ROTOR DOUBLE-SPEED INDUCTION MOTORS <i>George Todorov, Dimitar Jetchev</i>	220

46. USING OF THE DIAMAGNETIC PROPERTIES OF HTS FOR MAGNETIC SYSTEMS OF LINEAR ELECTRIC MACHINES <i>Ekaterina Kurbatova, Egor Kuschenko, Mikhail Sysoev, Andrey Drozdov, Pavel Dergachev, Pavel Kurbatov</i>	225
47. NEURAL CONTROL OF TWO-LINK PLANAR ROBOT <i>Zhivko Zhekov, Emil Marinov</i>	229
48. EXPLOITATION OF ADDITIONAL DEGREES OF FREEDOM OF MULTIPHASE DRIVES IN LOW FREQUENCY OPERATION OF MODULAR MULTILEVEL CONVERTERS <i>M. Daoud, A. Elserougi, A. Abdel-Khalik, A. Massoud, R. Bojoi, S. Ahmed</i>	233
49. AN APPROACH TO SOLVING RAMP START ISSUES IN SENSORLESS FIELD ORIENTED CONTROL WITH SLIDING MODE OBSERVER FOR PERMANENT MAGNET SYNCHRONOUS MOTORS <i>Emil Rachev, Vladislav Petrov</i>	238
50. STUDYING THE PERFORMANCE CHARACTERISTICS OF THE ELECTRIC VEHICLE <i>Marin Marinov, Javor Petrov, Orlin Stanchev, Pavel Andreev</i>	242
51. ANALYSIS OF FRACTIONAL-SLOT PERMANENT MAGNET SYNCHRONOUS MACHINES USED FOR HYBRID VEHICLES <i>Tsvetomir Stoyanov, Radoslav Spasov, Plamen Rizov</i>	247
52. EXPERIMENTAL SURVEY AND ANALYSIS ON THE IMPACT OF THE COMMUTATION PROCESS ON A POSITIVE DISPLACEMENT PUMP DRIVEN BY AN ASYNCHRONOUS ELECTRIC MOTOR <i>Peicho Tomov, Bohos Aprahamian, Krasimir Petrov, Pavel Andreev, Petko Petkov</i>	251
53. EXPERIMENTAL STUDY AND FUNCTIONAL REGIME ANALYSIS OF AN ELECTRIC PUMP UNIT WITH A THREE-PHASE ASYNCHRONOUS MOTOR <i>Peicho Tomov, Bohos Aprahamian, Georgi Petkov, Pavel Andreev, Petko Petkov</i>	255
54. MODELLING THE PROCESSES OF HYDRAULIC COMMUTATION IN A ELECTRO-HYDRAULIC PUMPING UNIT DRIVEN BY AN ASYNCHRONOUS MOTOR <i>Peicho Tomov, Dragiya Yanulov</i>	259
55. METHOD FOR INVESTIGATION OF THE PRESSURE VARIATION IN THE CHAMBERS OF GEAR PUMPS WITH SYMMETRIC AND ASYMMETRIC TOOTH PROFILES USED IN ELECTROHYDRAULIC DRIVE SYSTEMS <i>Wesam Mhana, Gencho Popov</i>	263
56. INVESTIGATION OF THE PRESSURE VARIATION IN THE CHAMBERS OF EXTERNAL GEAR PUMPS WITH SYMMETRIC AND ASYMMETRIC TOOTH PROFILES USED IN ELECTROHYDRAULIC DRIVE SYSTEMS <i>Wesam Mhana, Gencho Popov</i>	267
57. NEURO-FUZZY CONTROL SYNTHESIS FOR ELECTRO-HYDRAULIC FOLLOW-UP SYSTEM <i>Krasimir Ormandzhiev, Stanimir Yordanov</i>	271
58. STUDY OF PIEZOELECTRIC SPRING HARVESTERS <i>Nikola Georgiev, Raicho Raichev</i>	277
59. CRANKSHAFT POSITION MAGNETOELECTRIC SENSOR FOR CONTROLLER AREA NETWORK BUS <i>Roman Petrov, Ivan Belov, Mirza Bichurin, Slavcho Bozhkov, Ivan Milenov, Penko Bozhkov</i>	282

60. BAND-PASS ARC-FILTER WITH POLE FREQUENCY, POLE DECAY AND TRANSFER RATIO INDEPENDENT ADJUSTMENT <i>Darya Yu. Denisenko, Evgeny A. Zhebrun , Nikolay N. Prokopenko, Yuriy Iv. Ivanov</i>	286
61. INFORMATION SYSTEM FOR COLLECTION, PROCESSING AND PRESENTATION OF DATA FROM SENSOR NODES <i>Yordan Sivkov</i>	290
62. ANALYSIS OF FLUX DENSITY HARMONIC SPECTRUM OF THE COAXIAL MAGNETIC GEAR <i>Valentin Mateev, Miglenna Todorova, Iliana Marinova</i>	294
63. HARMONIC SPECTRUM FILTRATION FOR CURRENT SENSOR MEASUREMENTS <i>Dimitar Bogdanov, Martin Ralchev, Valentin Mateev, Iliana Marinova</i>	300
64. 3D COUPLED ELECTRIC-THERMAL-FLUID ANALYSIS OF BOLTED BUSBAR CONNECTION <i>Yanko Slavtchev, Valentin Mateev, Raina Tzeneva</i>	304
65. MODELING OF FLUID FLOW COOLING OF HIGH-SPEED ROTATIONAL ELECTRICAL DEVICES <i>Valentin Mateev, Georgi Ivanov, Iliana Marinova</i>	308
66. BIORTHOGONAL WAVELET FILTRATION OF SIGNALS USED IN THE INDUSTRIAL AUTOMATION SYSTEMS <i>Mariyana Todorova, Reneta Parvanova</i>	313
67. SERVO SYSTEM PARAMETER ESTIMATION BY WAVELET METHOD <i>Mariyana Todorova, Reneta Parvanova</i>	317
68. STUDY OF THE ELECTRICAL CHARACTERISTICS OF AUTOTRANSFORMER DISCRETE ALTERNATING VOLTAGE REGULATORS WITH R-L LOADS <i>Emil Panov, Milena Ivanova, Emil Barudov</i>	321
69. MULTIFUNCTIONAL ELECTRONIC TRANSFORMER <i>Aleksandr Antipin, Vladimir Udintsev, Sergey Sarapulov</i>	326
70. METHOD OF SOLVING THE PROBLEM OF THE REACTIVE POWER OF THE LOAD IN MULTILEVEL INVERTERS <i>Ekaterina Mirgorodskaya, Nikita Mityashin, Vadim Kolchev, Ivan Artyukhov, Sergei Stepanov</i>	331
71. UNIVERSAL POWER SOURCE OF SINGLE-PHASE MULTILEVEL INVERTERS <i>Ekaterina Mirgorodskaya, Nikita Mityashin, Vadim Kolchev, Yuri Tomashevsky, Sergei Stepanov, Denis Artyukhov</i>	337
72. NUMERICAL DESIGN OF COMPLEX SHAPE COILS <i>Dragomir Grozdanov, Nikolay Hinov</i>	342
73. COMPARISON BETWEEN CLASICAL AND MODERN METHODS FOR OPTIMAL CONTROL OF BUCK DC/DC CONVERTER <i>Nikolay Hinov, Gergana Vacheva, Bogdan Gilev</i>	346
74. MODELLING AND CONTROL OF BIDIRECTIONAL BUCK-BOOST CONVERTER FOR ELECTRIC VEHICLES APPLICATIONS <i>Gergana Vacheva, Vladimir Dimitrov, Nikolay Hinov</i>	350
75. MODELLING AND RESEARCH OF SYNCHRONOUS CONVERTER FOR ACTIVE BALANSING SYSTEM <i>Dimitar Arnaudov, Krasimir Kishkin</i>	354

76. FUELCELL POWER SUPPLY SYSTEM – MODELLING AND OPTIMIZATION <i>Vladimir Dimitrov, Plamen Punov, Dimitar Arnaudov</i>	358
77. ASPECTS OF MODERN COMPUTING OF MEDIUM VOLTAGE ELECTRICAL NETWORKS WITH NEUTRAL TREATED BY COMPENSATION COILS <i>Silvia-Maria Diga, Nicolae Diga, Adelaida-Mihaela Duinea, Cristian Bratu</i>	362
78. ANALYSIS OF THE OPERATIONAL PARAMETERS OF A MEDIUM VOLTAGE CABLE LINE <i>Dian Malamov, Ivan Hadzhiev, Vasilina Zlatanova, Albena Pavlova, Aleksandar Georgiev, Snejana Terzieva</i>	368
79. FORECASTING THE ENERGY CONSUMPTION IN AFGHANISTAN WITH THE ARIMA MODEL <i>Atanas Mitkov, Najmuddin Noorzad, Katerina Gabrovska-Evstatieva, Nicolay Mihailov</i>	373
80. METHOD OF STUDY OF THERMOELECTRIC GENERATORS <i>Kaloyan Ivanov, Anatoliy Aleksandrov</i>	377
81. MULTI-MODULAR WIND GENERATOR FOR URBAN WIND POWER <i>Ivan Artyukhov, Sergei Stepanov, Ekaterina Mirgorodskaya, Vasilina Kovalenko, Nikita Mityashin</i>	382
82. ELECTRIC POWER SUPPLY OF WIRELESS SENSORS BY THERMO-ELECTROCHEMICAL CELLS <i>Denis Artyukhov, Igor Burmistrov, Ivan Artyukhov</i>	388
83. ULTRACAPACITOR CHARGING FROM A SOURCE WITH VARYING AND LIMITED POWER <i>Hristo Antchev, Anton Andonov, Kostadin Milanov</i>	393
84. WEB-BASED ENVIRONMENT FOR VIRTUAL LABORATORIES IN THE FIELD OF ELECTRICAL ENGINEERING <i>Boris Evstatiev, Katerina Gabrovska-Evstatieva, Valentina Voynohovska, Ivan Beloev</i>	398
85. UNCERTAINTY CONSIDERATIONS OF MEASURING EQUIPMENT DURING EXPERIMENTAL RESEARCH IN MACHINING <i>Neven Trajchevski, Vasilija Sarac, Rose Smileski</i>	402
86. COMPARATIVE EVALUATION OF MECHANISMS FOR TRAFFIC PRIORITIZATION IN LTE NETWORKS <i>Aydan Haka, Veneta Aleksieva, Hristo Valchanov</i>	406
87. EXPLAINING THE DLT APPLICATIONS IN THE CONTEXT OF A CUSTOMERS, FACILITY MANAGEMENTS AND UTILITY COMPANIES RELATIONSHIP <i>Peyo Hristov, Teodora Hristova</i>	411
88. NEAR REGION ACOUSTIC DRIVER CIRCULAR CONE <i>Georgi Markov, Ekaterinoslav Sirakov</i>	416
89. COMPARATIVE ANALYSIS OF CONTENT-BASED IMAGE RETRIEVAL SYSTEMS <i>Miroslav Marinov, Irena Valova, Yordan Kalmukov</i>	420
90. INVESTIGATION OF ONLINE CORONA LOSSES AS A FUNCTION OF OPERATING VOLTAGE BASED ON PMU DATA <i>Anna Diachenko, Kirill Sidorov, Ammar Hussein, Yuri Kononov</i>	425
91. COMPARATIVE ANALYSIS OF HV CABLES ELECTRICAL CHARACTERISTICS <i>Dimitar Georgiev</i>	430

92. TRANSIENT STABILITY ANALYSIS IN ROTOR WINDING OF HYDROGENERATOR AT VARIOUS SHORT CIRCUIT VALUES IN POWER GRID IN CONSIDERATION WITH AEC <i>S.E. Kokin, M.KH Safaraliev, Kh.N. Rasulzoda, R.A. Otashbekov, J.B. Rahimov, Sh.M. Sultonov</i>	434
93. ELECTRICAL EQUIPMENT TECHNICAL STATE INTEGRAL ESTIMATE BASED ON ADDITIONAL NON-DESTRUCTIVE METHODS <i>Alexandra Khalyasmaa, Iana Nikitina, Lina Wang</i>	438
94. POWER SYSTEM SECTIONALIZING OPTIMIZATION USING GENETIC ALGORITHM <i>Albina Gavrilova, Alexandra Khalyasmaa</i>	443
95. THE DEVELOPMENT OF A JOINT MODELLING FRAMEWORK FOR OPERATIONAL FLEXIBILITY IN POWER SYSTEMS <i>Nikolai Voropai, Christian Rehtanz, Stefan Kippelt, Nikita Tomin, Ulf Haeger, Dmitry Efimov, Victor Kurbatsky, Irina Kolosok</i>	447
96. STATE OF HUMAN CAPITAL DEVELOPMENT AND WORKING CONDITIONS IN ELECTRICAL DISTRIBUTION COMPANIES IN BULGARIA <i>Kiril Anguelov</i>	453
97. OPPORTUNITIES FOR PREVENTION OF CORRUPTION IN PUBLIC PROCUREMENT IN ELECTRICITY DISTRIBUTION COMPANIES IN BULGARIA <i>Kiril Anguelov</i>	457
98. RESEARCH STUDIES FOR THE EFFICIENCY OF MANAGEMENT IN BULGARIAN ELECTRICITY DISTRIBUTION COMPANIES <i>Kiril Anguelov</i>	460
99. A MODEL FOR DEVELOPMENT OF PROFESSIONAL COMPETENCE IN STUDENTS OF TECHNICAL INSTITUTES OF HIGHER LEARNING <i>Diana Izvorska</i>	464
100. SOME PROBLEMS IN DISTANCE LEARNING ON THEORETICAL ELECTRICAL ENGINEERING IN TECHNICAL UNIVERSITY OF SOFIA <i>Simona Kirilova Filipova-Petrakieva</i>	469
101. A NOVEL METHOD FOR DETERMINING THE MAXIMUM ENERGY FLOWS, PASSED THROUGH THE ELECTRICAL POWER SYSTEMS UNDER INTERVAL UNCERTAINTY <i>Simona Kirilova Filipova-Petrakieva</i>	474
102. INNOVATION IN THE TEACHING METODOLOGY OF ELECTRONICS <i>Ilonka Lilyanova</i>	480
103. APPLICATION OF VARIABLE FREQUENCY DRIVES (VFD) WITH LARGE 6 KV ASYNCHRONOUS MOTORS <i>Valentin Kolev, Iva Draganova-Zlateva</i>	485
104. DETERMINATION OF THE THERMAL RESISTANCE OF ALUMINUM AND COPPER HEAT SINK IN A THERMOELECTRIC PUMP <i>Ivaylo Belovski, Anatoliy Aleksandrov</i>	488
105. IMPLEMENTATION OF COLLISION SENSE AND ORIENTATION SYSTEM <i>Stanislav Simeonov, Todor Kostadinov, Ivaylo Belovski</i>	492
106. BIOELECTRICITY PRODUCTION AND WASTE WATER TREATMENT BY AIR CATHODE MICROBIAL FUEL CELL <i>Blagovesta Midyurova</i>	496

107. PERFORMANCE ANALYSIS AND MODELLING OF GRID-CONNECTED SMALL PHOTOVOLTAIC SYSTEM <i>Atanas Chervenkov, Atanas Yanev, Todorka Chervenкова</i>	500
108. EXPERIMENTAL AND NUMERICAL ANALYSIS OF THE THERMAL FIELD IN A HIGH POWER LOW VOLTAGE FUSE <i>Ivan Hadzhiev, Dian Malamov, Iosko Balabozov, Ivan Yatchev</i>	504
109. COMPUTER MODELING AND EXPERIMENTAL VERIFICATION OF A HYBRID ELECTROMAGNETIC SYSTEM WITH MAGNETIC FLUX MODULATION <i>Ivan Yatchev, Iosko Balabozov, Hartmut Brauer, Vultchan Gueorgiev</i>	509
110. PARAMETRIZED MODELLING OF A HYBRID ELECTROMAGNETIC SYSTEM WITH MAGNETIC MODULATION <i>Ilona Iatcheva, Iosko Balabozov, Elena Yankulova, Ludmil Kirilov</i>	514
111. OPTIMAL ALGORITHMS FOR MINIMIZING OF ACTIVE AND REACTIVE POWER OF INDUCTION MOTORS TAKING INTO ACCOUNT OF IRON LOSSES <i>Dimitar Spirov</i>	518
112. SIMULATION FRAMEWORK FOR BUILDING OF 6LOWPAN NETWORK <i>Aydan Haka, Rusen Vasilev, Veneta Aleksieva, Hristo Valchanov</i>	522
113. AN APPROACH FOR HOST BASED BOTNET DETECTION SYSTEM <i>Yulia Aleksieva, Hristo Valchanov, Veneta Aleksieva</i>	527
114. COMPARATIVE ANALYSIS OF PROTOTYPES BASED ON LI-FI TECHNOLOGY <i>Diyana Dinev, Veneta Aleksieva, Hristo Valchanov</i>	531
115. DETERMINATION OF THE SYNCHRONOUS INDUCTANCES OF A CLAW POLE ALTERNATOR <i>Vladimir Lazarov, Zahari Zarkov, Ivan Bachev</i>	535
116. EXPERIMENTAL STUDY OF ELECTRONICALLY CONTROLLED CLOSED-LOOP ELECTROHYDRAULIC SYSTEM WITH VARIABLE DISPLACEMENT PUMP <i>Ilcho Angelov, Nikola Stanchev, Alexander Mitov</i>	541
117. FACTOR ANALYSIS OF THE QUALITY OF ELECTRICITY <i>Vania Rangelova, Vasil Drambalov</i>	546
118. ASSESSMENT OF CONDITIONS FOR THE APPLICATIONS OF DLT FOR SMART METERING IN BULGARIA ACCORDING TO THE EUROPEAN REQUIREMENTS <i>Teodora Hristova, Peyo Hristov</i>	551
119. MODULE FOR TAXIMETER USE OF CAN INTERFACE IN CAR <i>Nikolay Valov, Irena Valova, Donka Ivanova, Daniel Chervenski</i>	557
120. ROBUST CONTROL APPLICATION POSSIBILITIES FOR MACHINING ROBOTS <i>Sasho Guergov, Nina Nikolova, Grigor Stambolov</i>	562
121. PID CONTROL OF ANALOG THREE-DIMENSIONAL PLANT <i>Bozhidar Rakov, Georgi Ruzhekov</i>	566
122. REAL TIME MONITORING OF HV ELECTRICAL MACHINES DRIVING POWERFUL PUMP AGGREGATES <i>Vultchan Gueorgiev, Plamen Rizov</i>	571
123. WEARABLE DEVICE INTERFACE COMPARISON <i>Yuri Dimitrov, Veneta Aleksieva, Hristo Valchanov</i>	575
124. SIZING OF PV BASED POWER SUPPLY FOR IRRIGATION SYSTEM – APPLICATION IN SANDANSKI, BULGARIA <i>Ludmil Stoyanov, Ivan Govedarski, Vladimir Lazarov</i>	579

125.	FORECASTING THE PRODUCTION OF ELECTRICAL ENERGY FROM PHOTOVOLTAICS <i>Marya Marinova, Bohos Aprahamian, Tatyana Dimova, Pavel Andreev</i>	585
126.	COMPARISON OF WIND DATA MEASUREMENT RESULTS OF 3D ULTRASONIC ANEMOMETERS AND CALIBRATED CUP ANEMOMETERS MOUNTED ON A MET MAST <i>Dmitriy Dubov, Bohos Aprahamian, Meline Aprahamian</i>	590
127.	REGRESSION ANALYSIS OF EXPERIMENTAL DATA FOR ELECTRICAL PARAMETERS OF WATER AT A HIGH VOLTAGE PULSE DISCHARGE <i>Milena Ivanova</i>	595
128.	A STUDY OF ANODIZING OF TI-6AL-7NB ALLOY <i>Ivaylo Nedelchev, Deyan Veselinov, Hristo Skulev</i>	600
129.	SIGNAL PROCESSING AT ANODIZING OF TI-6AL-7NB ALLOY <i>Ivaylo Nedelchev</i>	606
130.	SOME ASPECTS OF LASER CUTTING OF THICK-WALLED WORKPIECES FROM PHASE-SHIFTING STEELS <i>Vladimir Shtarbakov, Maik Streblau</i>	611
131.	ON DETERMINING THE ENERGY EFFICIENCY OF THE ELECTRIC OVEN <i>Yanita Slavova, Mariya Marinova</i>	615
132.	RESEARCH OF THE MAGNETIC FIELD INSIDE A DRUM SEPARATOR WITH PERMANENT MAGNETS <i>Tatyana Dimova, Bohos Aprahamian and Mariya Marinova</i>	621
133.	ELECTROMAGNETIC PROCESSES IN A THREE-PHASE INDUCTION DEVICE UNDER LOAD <i>Georgi Zhelev, Marin Marinov, Maik Streblau</i>	625
134.	A MODEL FOR SIMULATION OF NONLINEAR INDUCTORS IN VIRTUAL ENVIRONMENTS <i>Boris Evstatiev, Dimcho Kiriakov, Dimitar Trifonov</i>	629
135.	DEVELOPMENT OF A VIRTUAL LABORATORY IN LOGIC DESIGN <i>Boris Evstatiev, Nadezhda Evstatieva, Dimitar Trifonov</i>	633
136.	DENOISING OF PULSAR SIGNAL USING WAVELET TRANSFORM <i>Ivan Garvanov, Ruska Iyinbor, Magdalena Garvanova, Nikolay Geshev</i>	637
137.	ONE APPROACH TO IMPROVE THE SOFTWARE QUALITY BY APPLYING SOFTWARE DESIGN PATTERNS <i>Dimitrichka Nikolaeva, Violeta Bozhikova</i>	641
138.	A METHOD OF BINARY COMPRESSION <i>Gergana Spasova</i>	647
139.	REDUCED DATABASE FOR VOICE COMMANDS RECOGNITION USING CLOUD TECHNOLOGIES, ARTIFICIAL INTELLIGENCE AND DEEP LEARNING <i>Snezhana Pleshkova, Aleksander Bekyarski, Zahari Zahariev</i>	652
140.	QR CODE'S MAXIMUM SCANNING DISTANCE INVESTIGATION <i>Pavel Andreev, Bohos Aprahamian, Marin Marinov</i>	656
141.	AN APPROACH FOR ANALYSIS OF THE MAGNETIC FIELD OF INDUCTION MOTOR WITH A SQUIRREL-CAGE ROTOR <i>Marin Todorov, Marin Marinov, Maik Streblau</i>	660
142.	SENTINEL-1 SAR SLC IMAGE SHIP DETECTION AND MOTION ESTIMATION <i>Nikolai Kolev, Miroslav Tsvetkov, Chavdar Alexandrov</i>	664

143.	SENTINEL - 1 SAR IMAGE PROCESSING FOR TARGET DETECTION AND EVALUATION BY USING BULGARIAN VTMS <i>Chavdar Alexandrov, Miroslav Tsvetkov, Nikolay Kolev, Yordan Sivkov, Avgustin Hristov</i>	670
144.	COMPARISON BETWEEN INFORMATION PROVIDED BY RADAR AND AIS IN THE INTEGRATED VESSEL TRAFFIC SYSTEMS <i>Aleksandrina Angelova, Chavdar Alexandrov</i>	676
145.	LIFE DATA ANALYSIS AND OPERATIONAL RELIABILITY POINT ESTIMATION RELATED TO MEDICAL ELECTRONIC DEVICES <i>Julia Garipova, Anton Georgiev, Toncho Papanchev, Dimitar Zlatev</i>	681
146.	AN APPROACH TO STORING REAL-TIME SENSOR READINGS BASED ON NOSQL DATABASE SYSTEMS <i>Georgi Georgiev, Milko Marinov</i>	686
147.	MICROPROCESSOR BASED ELECTRONIC SYSTEM FOR CONTROL OF INDUCTION HEATING IN MACHINE INDUSTRY <i>Seher Kadirova, Ivan Evstatiev, Daniel Kajtsanov, Teodor Nenov</i>	691

PID Control of Analog Three-Dimensional Plant

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Abstract - Multi-loop (decentralized) Proportional-Integral-Derivative controllers find extensive use in industrial systems. An auto-tune method is present in the paper. The independent design approach is used. The method treats the multivariable system as independent SISO systems with interactions as an additive uncertainty. Modified relay feedback is applied. A software is developed using Siemens TIA Portal V15. The approach is tested on a three-dimensional multivariable system, modelled by an analog board.

Keywords— Auto-tune, Multi-loop PID, multivariable system, PLC

I. INTRODUCTION

Despite the existence of advanced methods and the big research around them, the PID control is still dominant in industrial systems. The main factors are simple structure, small number of tuning parameters, possibility for fine manual tuning. Most of the tuning methods are based on simple models, making auto tuning easier task. For a long time there exist software implementations and libraries of the algorithm. All of this makes the practical implementation easier.

The main problem when dealing with MIMO systems is the existence of interconnections. Reference or parameters change, presence of disturbance or noise in one of the loops impact all the others and because of the feedback the last form a reaction to the source loop. Treating this problem is of utmost importance when controlling MIMO plants. In the literature one can find different methods like – full dynamic decoupling [1], decoupling for fixed frequency range [2] or static decoupling, effective open loop [3], sequential tuning, detuning, inverted decoupling [4]. The present work use the principle of independent design [5]. A control system for PID auto tuning is developed using minimal plant information.

II. INDEPENDENT DESIGN

The essence of the independent design is finding n independent controllers in the presence of some constraints imposed by the interconnections of the MIMO plant. Two problems arise, first the bigger the interconnection strength the bigger the constraints imposed are, the poorer the performance of the closed loop system would be. Second, the lack of information for other loop’s controllers additionally could degrade performance. The present work offers an approach for auto tuning with minimal information.

Let the plant is described by its transfer matrix (1), where m is the number of inputs, r is the number of outputs. For a decentralized control a square matrix is required, i.e. m=r.

$$G(s) = \begin{bmatrix} G_{11}(s) & \dots & G_{1m}(s) \\ \dots & \dots & \dots \\ G_{r1}(s) & \dots & G_{rm}(s) \end{bmatrix} \quad (1)$$

For the MIMO plant to be treated as m SISO plants let’s assume that each diagonal element of the transfer matrix is a nominal plant of an uncertain family of plants. Then one assumes that the sum of each row of the off-diagonal elements represents additive uncertainty for each family (2).

$$\left| \begin{array}{l} G_p^i(s) = G_n^i(s) + G_a^i(s) \\ G_n^i(s) = G_{sh}^{ii}(s) \\ G_a^i(s) = \sum_{i=1, j=1}^{r, m} G_{sh}^{ij}(s), i \neq j \end{array} \right. \quad (2)$$

Where $G_p^i(s)$ is the uncertain (or perturbed) plant, $G_n^i(s)$ is the nominal plant, $G_a^i(s)$ is the additive uncertainty. Superscript i used in the first equation of (1) changes from 1 to m and is used to enumerate the uncertain, nominal plant and the accompanying uncertainty for every loop. $G_{sh}^{ii}(s)$ represents the i -th diagonal element of the transfer matrix $G_{sh}^{ii}(s)$ of the plant. $G_{sh}^{ij}(s)$ represents all the off-diagonal elements of the transfer matrix $G_{sh}(s)$. Treating the interconnections as additive uncertainty could lead to conservative results regarding system performance, but this approach makes the task easier and possible to use approved SISO methods with small modifications.

For every model $G_n^i(s)$ one could find a PID controller, providing stability margins in the presence of the additive uncertainty $G_a^i(s)$. One easy approach is using the maximum of the sensitivity function S . The reciprocal point of the maximum is the smallest distance to the Nyquist point, Fig. 1.

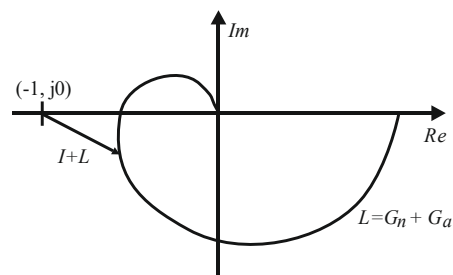


Fig. 1. Nyquist characteristic

By using the second method of Astrom and Hagglund [6] (the kappa method) for every $G_p^i(s)$, one finds a PID controller ensuring that the nominal plant $G_n^i(s)$ does not encircle the Nyquist point in the presence of an additive uncertainty $G_a^i(s)$. To do that the experiment using relay feedback should be modified. For every diagonal element $G_n^i(s)$ the feedback should be closed only by output y_i , but the relay output should be fed to all of the plant inputs (Fig. 2.), leading to an experiment using $G_p^i(s)=G_n^i(s) + G_a^i(s)$.

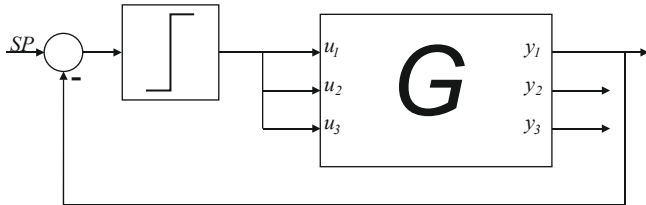


Fig. 2. Modified scheme for relay auto-tuning

One should note that this approach is feasible only for small interconnections. To address this issue one should make a pairing analysis using a tool like RGA, RNGA and the Niederlinski theorem. Some kind of compensation is also possible. Some authors suggest using the following [7]:

$$G_{sh}(s) = G(s).D, \quad (3)$$

$$D = G^{-1}(0) \quad (4)$$

D is a constant matrix, ensuring the corrected plant G_{sh} is an identity matrix in steady state. This removes the influence between the loops at steady state at the expense of increased high frequency interconnections. To address the issue one could use a second order filter [8] at the output of every PID, ensuring high roll-off, leading to some kind of robustness.

III. THE PLANT

The plant to be *controlled* consists of three inputs and three outputs. A general structure of the plant is shown on Fig. 3.

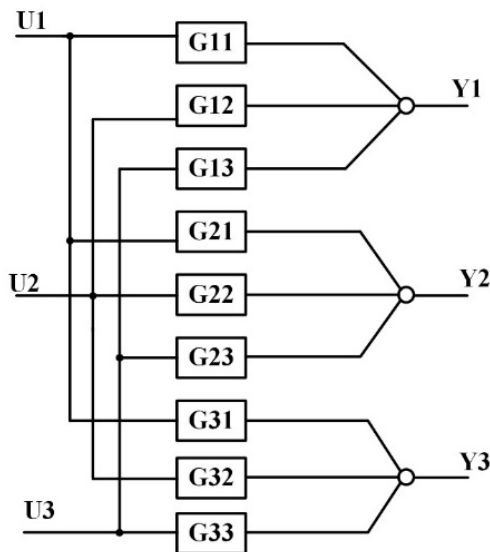


Fig. 3. Plant structure

The diagonal elements of the object are second order dynamics, the off-diagonal elements are of first order. The interconnections have strong influence, especially in steady state. There is also dynamics with opposite directions. The control object is modeled by an analog board “UNIVERSAL BOARD 2”. First order dynamics, proportional gains and sum blocks are used. The plant modeled on the analog board is shown on Fig. 4. The parameters of the building blocks are set manually. Because of the poor accuracy of the potentiometers, one needs to evaluate every block’s parameters by experiment. Each block’s input is fed voltage by the DACs of the PLC, the outputs of the blocks are fed to the ADCs. A transient response is recorded, proportional gain and time constant of every block are estimated. The parameters are shown on TABLE I, where T_i represent time constant, K represent proportional gain. Picture of the PLC is shown on Fig. 5.

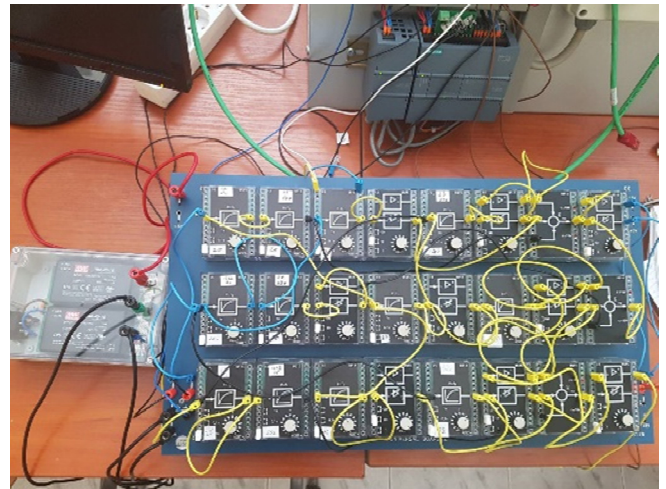


Fig. 4. Plant model on the analog board

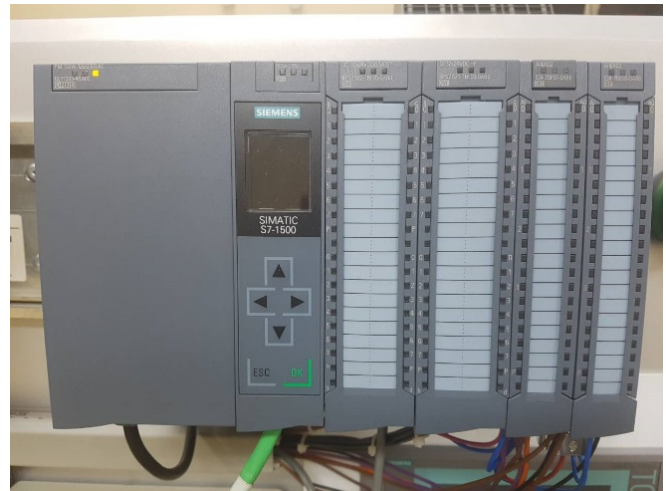


Fig. 5. PLC – Siemens CPU S7-1500

IV. SOFTWARE

The software is developed using TIA Portal V15. The program doing all the computations is written, using SCL. For ease of use HMI station is created. A structure of the closed loop system is shown on figure 11. In general the control algorithm consists of two modules. The first module contains the three discrete PIDs [9] - (5), (6). The second module contains the decoupling matrix (4).

$$u(k) = K_p \left(e(k) + I(k) + \frac{T_d}{T_0} (e(k) - e(k-1)) \right) \quad (5)$$

$$I(k) = I(k-1) + \frac{T_0}{T_i} e(k), \quad (6)$$

$$I(0) = 0.$$

TABLE I. Plant parameters

Block notation	T_1, s	T_2, s	K
G_{11}	11.5	16.6	1
G_{12}	4	-	-9
G_{13}	34	-	8
G_{21}	19	-	-5
G_{22}	7	33.0	8
G_{23}	3	-	7
G_{31}	5	-	-9
G_{32}	14	-	3
G_{33}	15	24.8	1

Where u is the output of the PID, e is the control error, k is the current step, K_p is the proportional gain, T_i is the integral time constant, T_d is the differential time constant of the controller, T_0 is the sample time, chosen as 0.1s.

The HMI makes it possible to estimate the proportional gains of every element of the transfer matrix, calculate the compensation matrix D , auto tune the PIDs, choose control mode (auto/manual), observe closed loop processes like control action and process variables. Some of the functionality of the HMI is shown on Fig. 6, Fig. 7, Fig. 8 and Fig. 9.

	U	G			Y
Calc 1	+1.00	+0.955	-8.269	+7.968	+0.955
Calc 2	+0.00	-4.503	+7.292	+6.121	-4.506
Calc 3	+0.00	-8.695	+2.912	+0.969	-8.697

Fig. 6. Calculating gains

	D
Calc D	-0.014 +0.040 -0.138
Det	-0.061 +0.088 -0.052
	+0.063 +0.087 -0.037

Fig. 7. Calculating compensation matrix D

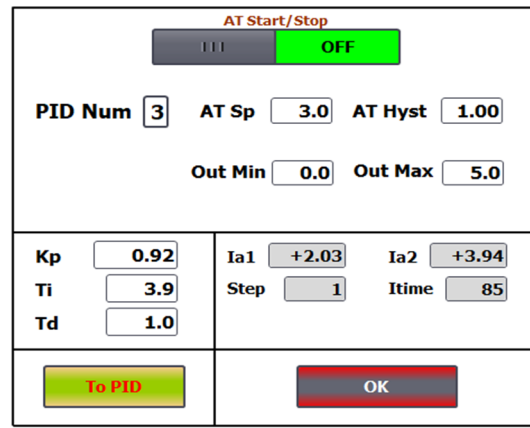


Fig. 8. Auto tuning panel

All	1	2	3
Auto/Manual	Start/Stop	Start/Stop	Start/Stop
Auto/Man	Auto/Man	Auto/Man	Auto/Man
SP	+3.00	+3.00	+3.00
ManOut	+3.07	+3.05	+3.02
Kp	+6.72	+3.87	+3.91
Ti	+1.69	+1.06	+1.40
Td	+0.40	+0.26	+0.34
OutMax	+10.00	+10.00	+10.00
OutMin	-10.00	-10.00	-10.00

Fig. 9. PIDs control panel

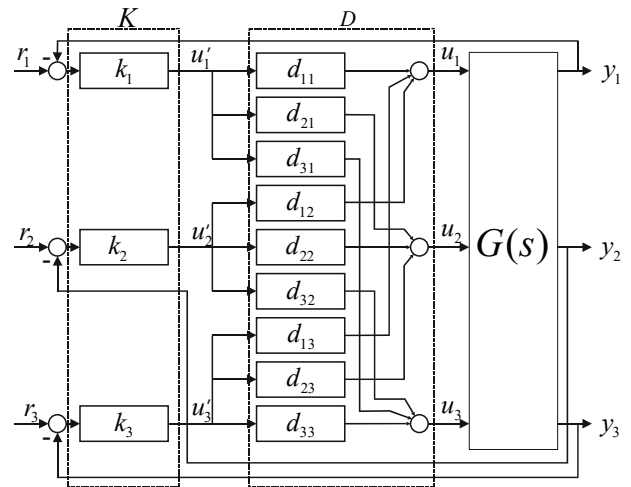


Fig. 10. Structure of the closed loop system

V. RESULTS

For the purpose of the present work it is assumed that the model of the MIMO system is unknown. First step is estimating the proportional gains of every element of the transfer matrix.

Sequentially every input is fed constant voltage, the steady state values of the outputs are recorded. Using the newly found steady state matrix one calculates the compensation matrix D for the second step. Next using modified relay feedback as shown on figure 5, one finds the three PID controller parameters. Result of the relay feedback for the first loop is shown on Fig 11.

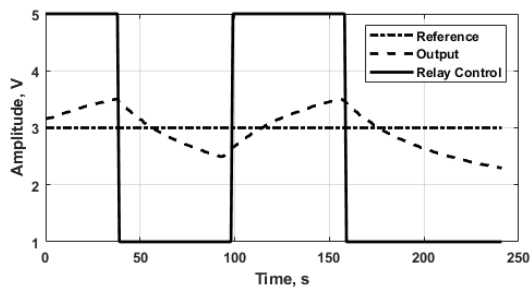


Fig. 11. Relay feedback for loop one

A symmetrical oscillation around the reference is observed, although it is not too close to a sinusoidal one. The form is dictated by the dynamics of the plant. Estimating the ultimate parameters (K_u – ultimate gain, T_u – ultimate period) after every loop’s experiment, one could calculate the parameters of every PID. When the parameters of the three PIDs are found one should check the closed loop system performance. Sequentially every input is fed 3V then 0V, the results are shown on Fig. 12, Fig13 and Fig. 14.

All loops are characterized by fast transients. Loop two and three exhibit overshoot. The interconnections are either small or fast decaying. One important observation is the difference of the transient responses going up or down. Although the experiments are conducted using signal in the linear range of the board, there is still some nonlinearities. Summary of the performance of the closed loop system could be found in table two and table three. TABLE II contains transient response characteristics going up and down. TABLE III contains interconnection estimation by *ISE (Integral of the squared error)* criterion.

TABLE II. Time domain performance of the closed loop system

	Rise Time [s]	Settling time [s]	Overshoot %	Rise Time [s]	Settling time [s]	Overshoot %
Loop 1	2.7	4.8416	0	1.857	4.8482	0
Loop 2	1.5193	3.7945	12.2685	1.2653	3.689	23.05
Loop 3	2.1384	5.5951	7.0602	1.5756	5.7257	12.1528

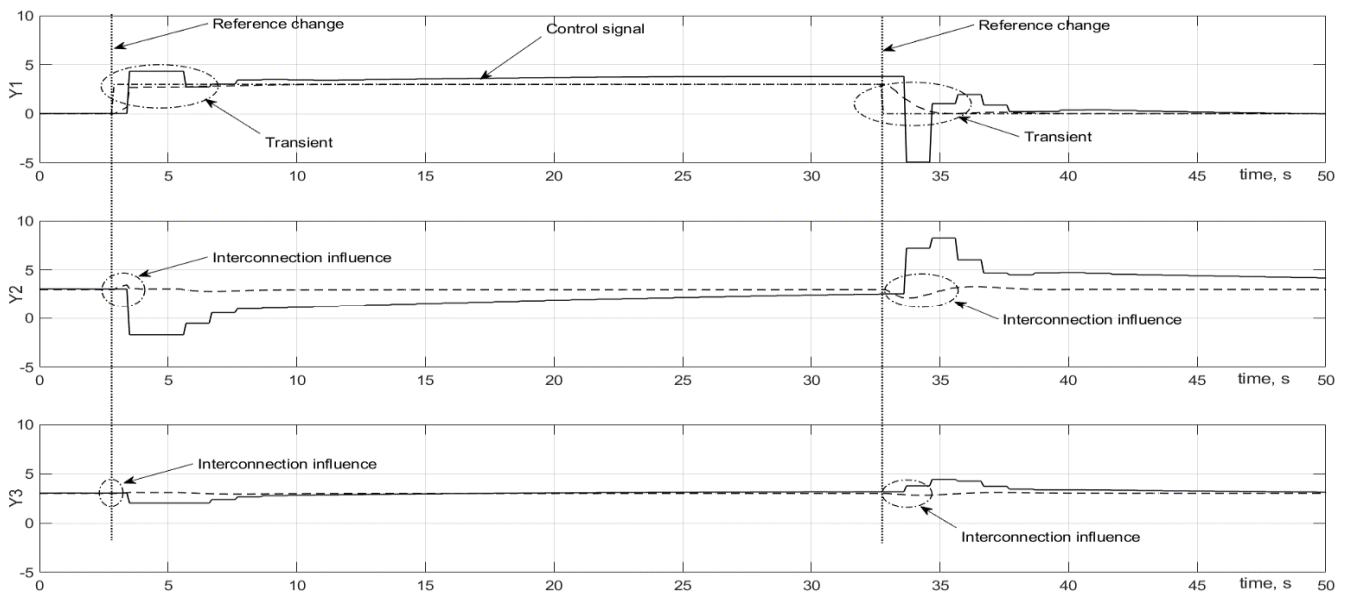


Fig. 12. Reference change at input one.

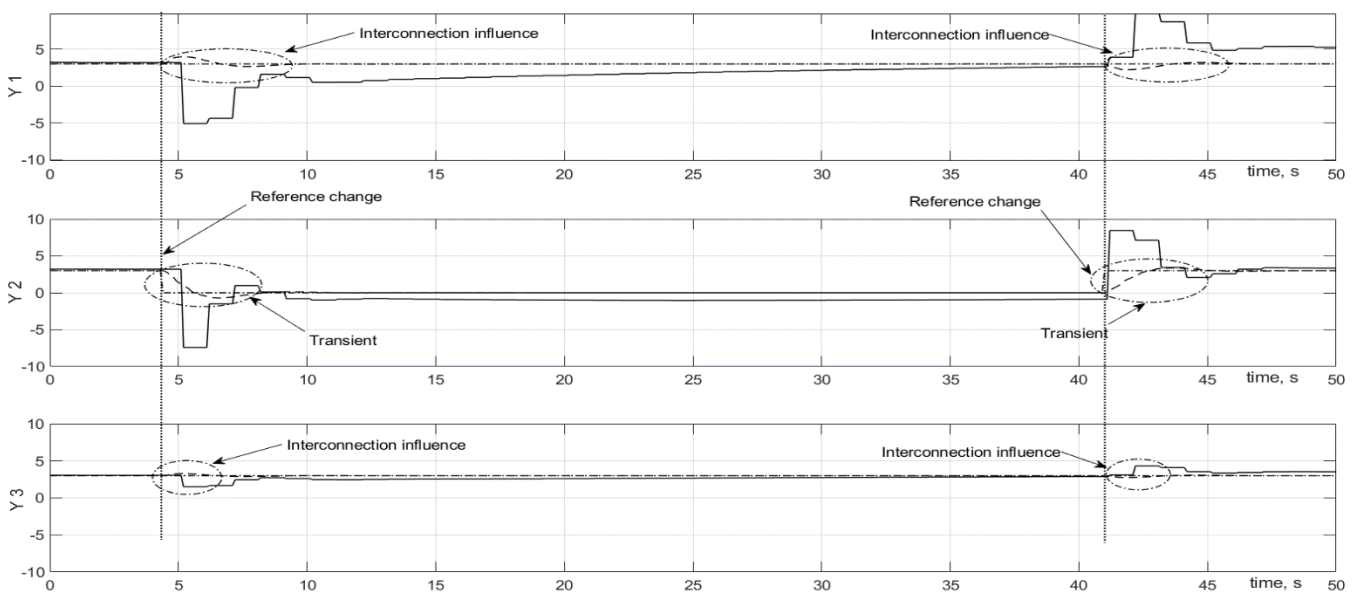


Fig. 13. Reference change at input two.

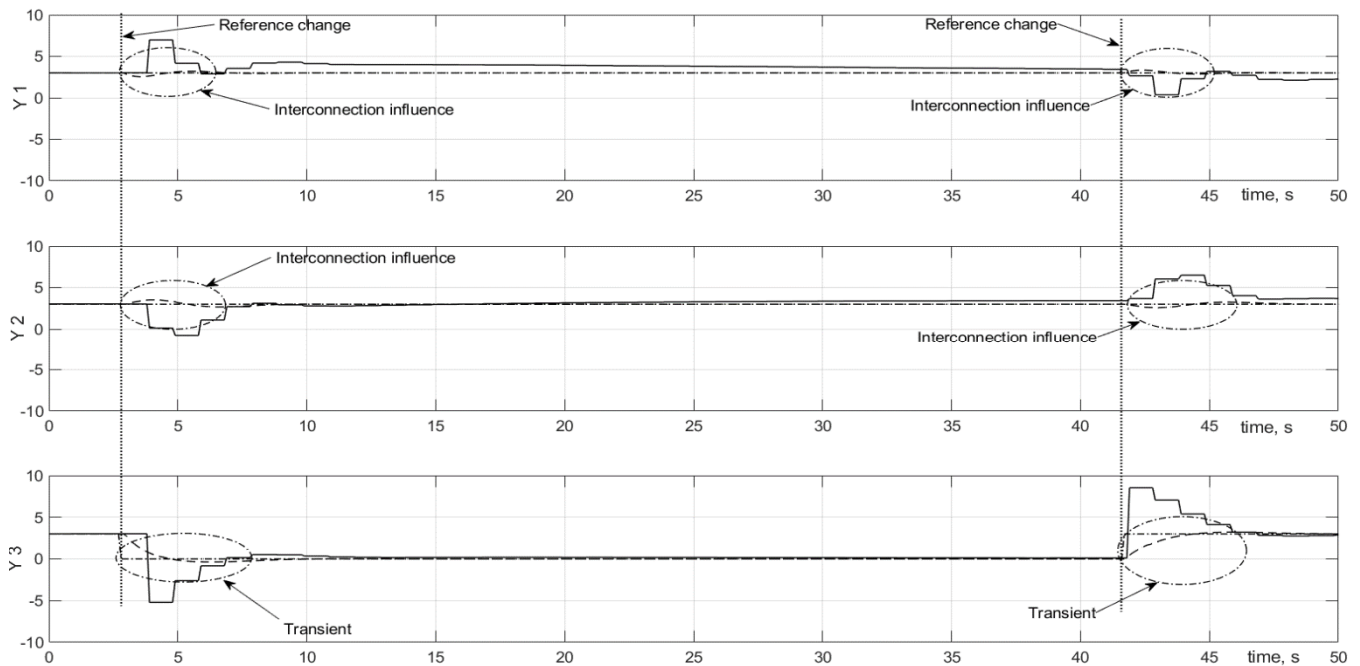


Fig. 14. Reference change at input two.

TABLE III. An interaction measure by ISE criterion

	J_{u1}	J_{u2}	J_{u3}	J_{d1}	J_{d2}	J_{d3}
Loop 1	-	8.5416	1.3308	-	11.9201	2.1888
Loop 2	1.3958	-	3.3843	9.3892	-	5.0739
Loop 3	0.2634	0.7553	-	0.6211	1.0247	-

VI. CONCLUSION

This paper presents an auto-tune method for MIMO plant, using only information about the steady state. An optional step would be some kind of simple plant shaping or decoupling procedure in the presence of strong coupling. The present work employs a static decoupling. The remaining dynamics of the interconnections are treated as additive uncertainty. The ideology of the method is based on the independent design. The plant is to be controlled by PLC. All the necessary programs are developed using Siemens TIA Portal V15. The plant is modelled by an analog board. Three PID controllers are tuned using a modified relay experiment, making it possible to include the additive uncertainty into the procedure. Some experiments are conducted with the closed loop system, consisting of three independent PID controllers, static decoupling matrix and the plant. The step response of every loop is fast with as little as possible overshoot. The influence of the interconnections is small, thanks to the compensation matrix and the inclusion of the interconnections in the design procedure. The overall quality of the system is good, as is observed by the figures and the tables presented. One should note that the present approach is feasible when the influence of other loops is not as large or could be reduced by simple shaping or decoupling, otherwise the tuned parameters of the PID controllers would be too conservative, leading to poor quality of the closed loop response. Another weak point is the omission of a real uncertainty (not the artificial one, introduced by the interconnections) from the design procedure. Although that point is less of a concern, having in mind that one uses real experiment with the real plant and the

fact that the procedure is simple, making it possible to retune the PIDs, if the plant parameters change notably.

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